

# E-GOLDradio

GSM/GPRS Single Chip Solution

PMB 7870

From V2.1F to V2.2x

**CONFIDENTIAL**  
*Distribution with NDA by Marketing*

Secure Mobile Solutions



N e v e r   s t o p   t h i n k i n g .

**Edition 2005-07-19**

**Published by Infineon Technologies AG,  
St.-Martin-Strasse 53,  
81669 München, Germany**

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Never stop thinking.

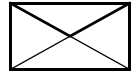
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# 1 E-GOLDradio Hardware Issues

**E-GOLDradio**
**CONFIDENTIAL**
**Revision History:** 2005-07-19

Rev. 1.02

Previous Version: 1.01, 2005-07-15

Page Subjects (major changes since last revision)

Changes for Rev. 1.01

	Removed UTP Issue: WS00005882: <b>I2S: Timing for Master Mode Not Correct</b>
	Added UTP Issues: <b>WS00005901</b> , <b>WS00006687</b> , <b>WS00007440</b> , <b>WS00007736</b> , <b>WS00008364</b> , <b>WS00008373</b> , <b>WS00008476</b> , <b>WS00009022</b> , and <b>WS00009028</b>

Changes for Rev. 1.02

	UTP issues solved in E-GOLDradio from V2.1F to V2.2x: <b>WS00006687</b> , <b>WS00007440</b> , <b>WS00007600</b> , <b>WS00007724</b> , <b>WS00007736</b> , <b>WS00008085</b> , <b>WS00008097</b> , <b>WS00008142</b> , <b>WS00008373</b> , <b>WS00008476</b>
	Added UTP Issue: <b>WS00008988</b>

## 1.1 Overview of Hardware Issues

**Table 1-1 Overview of HW Issues**

Type	Title	ID
DSP	<b>BBRX: The Adjacent Channel Power Estimation Is Biased By Transients</b>	WS00005865
DSP	<b>AFE: Noise Signals At Earpiece Buffer Output During Buffer Power Down</b>	WS00005868
DSP	<b>I2S: WA0/1 Signal Is Running While TX/RX Is Not Started</b>	WS00005876
DSP	<b>I2S2: RX Path Buggy (polarity bit = 0, Normal Master Mode in 6-Pin Config)</b>	WS00005877
DSP	<b>I2S: RX PCM Does Not Work After Restart</b>	WS00005878
DSP	<b>I2S Doesn't Always Work in Burst Mode</b>	WS00005880
MCU	<b>SIM: SIM_IN IRQ Illegally Occurs After Reset</b>	WS00005889
MCU	<b>SIM: 'Ghost Character' after Retransmission at Parity Error</b>	WS00005893
DSP	<b>SSC DSP: Interrupt Bug during Transparent Mode</b>	WS00005897
MCU	<b>Corruption of PEC src/dst Segment Address When Overridden by Other PEC/INT</b>	WS00005899
MCU	<b>Dummy RTC_INT/RTC_T14 Interrupts After Reset Phase</b>	WS00005900
MCU	<b>RTC: Power Consumption and Current Peak</b>	WS00005901
MCU	<b>PM_INT: the Associated Interrupt Is Missing</b>	WS00005903
MCU	<b>EBU: Not Enough GPIO Can Be Used in VDD_EBU Domain (Mainly for WP of Flash)</b>	WS00005905
Top Level	<b>I2S: Burst Mode Not Capable Of Quasi-Continuous Transmission</b>	WS00005909
MCU	<b>SIM: Auto Powerdown Not SIM Standard Compliant</b>	WS00005911
DSP	<b>RTC: Alarm Interrupt Generation</b>	WS00005913
MCU	<b>CR110651: Delayed ILLOPA Trap Execution If Caused by Double-Indirect MOV Instruction</b>	WS00005918
MCU	<b>CR110652: Delayed ILLBUS Trap Execution When Writing External Bus</b>	WS00005919
MCU	<b>CR110653: Visible Mode Limitation for Read Accesses on X-Bus</b>	WS00005920
MCU	<b>CR110654: DIP Shows Incorrect Value If Breakpoint Is Inside Atomic/External Sequence</b>	WS00005921

**Table 1-1 Overview of HW Issues**

Type	Title	ID
MCU	<b>CR110655: Break-after-Make Does Not Work Correctly in Atomic Scenarios</b>	WS00005922
MCU	<b>CR110656: Read Address Triggers Cannot Be Generated for All Instruction Reads</b>	WS00005923
MCU	<b>CR110657: ILLBUS Exception Inside ILLBUS ISR Problem</b>	WS00005924
MCU	<b>CR110764: External Break Events (BREAK Pins) Can Be Lost</b>	WS00005925
MCU	<b>CR111192: Bit-Manipulating Instructions on EXICON/EXISEL Write Wrong Values</b>	WS00005926
MCU	<b>CR112452: Wrong SP Used with SCXT Reg, Mem, and Mem Pointing to SP</b>	WS00005927
Backend	<b>RTC: Internal Oscillator Instability</b>	WS00006183
DSP	<b>TEAKlite: Loop (rep) Ends Unexpectedly After Interrupt in PROM Page 1</b>	WS00006189
Top Level	<b>DSP: Consecutive Write Operation to Some Registers Fails</b>	WS00006272
MCU	<b>Boot Code: Access in External Flash during the BOOTROM</b>	WS00007413
Physical Implementation	<b>EBU: Improve Timings for EBU Interface</b>	WS00007486
MCU	<b>GSM Timer Unit: TFSKIP.SKIPC Bit Has No Effect</b>	WS00007537
MCU	<b>SCCU X_BUS Clock at 78 MHz</b>	WS00007621
Evaluation Board	<b>SSC: Phase Error When High Baudrate Is Set</b>	WS00008268
Verification	<b>Cannot Make Write Access to Internal Memory Immediately After Resetting MST_CLK_CTRL.CPUH</b>	WS00008364
Top Level	<b>CS2 Usage Restriction</b>	WS00008988
Verification	<b>C166S: Jump-Bit Executed Wrongly after Byte Manipulation</b>	WS00009022
Verification	<b>CGU: Phaseshifter Startup Problems Under Certain Circumstances</b>	WS00009028



## **1.2 Details of Hardware Issues**

### **1.2.1 CS2 Usage Restriction**

**ID**

WS00008988

**Type**

Top Level

**Description**

The external CS2 signal is an alternate function of the pad OE.

After reset this pad has the behavior of the OE function, that means if a device is connected on this pad for CS2, there may be a conflict on the data bus if another device (Flash) is connected on CS0. During the boot, when an access to CS0 is done to check if a code is in the external Flash, the OE pin (CS2) is also active and the CS2 device can drive the data bus at the same time as the CS0 flash.

Because the page mode switch is not supported by E-GOLDradio V2.1F to V2.2x, the OE function is not needed and the RD pad can be connected to the OE Flash input for CS0.

**Solution**

Do not use CS2.

## **1.2.2 CGU: Phaseshifter Startup Problems Under Certain Circumstances**

**ID**

WS00009028

**Type**

Verification

**Description**

It has been observed that a phaseshifter output clock may be missing immediately after switching on and removing the phaseshifter bypass for some variable time.

The problem has been seen with higher probability for a small voltage band (of around 40 mV) of the PLL + Core power supply at a given temperature.

If the missing clock is driving an active subsystem, the subsystem stops until the phaseshifter clock is started.

On E-GOLDRadio, the problem has been observed only with the phaseshifter1. For phaseshifter2, Phs2\_X only uses 0 and 2, so the problem has not been observed here.

System impact is low (when DSP clock is set to 78, 113.5 or 124.8 MHz).

**Solution**

1. Do not use values of Phs\_X = 4 and Phs\_X = 5 for phaseshifter2.
2. If a phaseshifter frequencies with Phs\_X = 4 (or 5) is necessary in the system, the phaseshifter must be started using the following procedure ("switching on the fly"):
  - a) Set PHX\_CTRL.x to 3
  - b) Power on the phaseshifter
  - c) Wait for 1  $\mu$ s
  - d) Switch on-the-fly towards the needed settings (for example, PHX\_CTRL.x = 4)
  - e) Wait for 1  $\mu$ s
  - f) Activate phaseshifter clock by removing bypass.

### **1.2.3 C166S: Jump-Bit Executed Wrongly after Byte Manipulation**

#### **ID**

WS00009022

#### **Type**

Verification

#### **Description**

AI00025116: Jump-bit wrongly executed after Byte manipulation instructions

When a JB/JNB/JBC/JNBS on a GPR follows an instruction that performs a byte write operation (MOVB, ADDB/ADDBC, ANDB, XORB, ORB, NEGB, CPLB, SUBB/SUBCB) on the same GPR but on the byte that is not been used by the JB/JNB/JBC/JNBS (that is, the byte where the bit used by the Jump is not located), the program flow gets corrupted. That means, the Jump may be wrong (or not done).

A side effect of this bug is that further program branches may also lead to illegal instruction fetches (fetches are performed from wrong addresses).

- Example 1

; Assume Rx.0 is 0 (x any GPR 0..7)

MOVB RxH, any\_value ; Any Byte write instruction on RxH

JB Rx.0, jump\_address ; WILL BE INCORRECTLY TAKEN

- Example 2

; Assume Rx.y is 1 (x any GPR 0..7; y any bit except bit0)

MOVB RxL, any\_value ; Any Byte write instruction on RxL for y= 8..15 or RxH for y= 1..7

JNB Rx.y, jump\_address ; WILL BE INCORRECTLY TAKEN

- Example 3

; Assume Rx.0 is 0(x any GPR 0..7)

MOVB RxH, any\_value ; Any Byte write instruction on R0H

JNB Rx.0, jump\_address ; WILL BE INCORRECTLY TAKEN

- Example 4

; Assume Rx.y is 1(x any GPR 0..7; y any bit except bit0)

MOVB RxL, any\_value ; Any Byte write instruction on RxL for y= 8..15 or RxH for y= 1..7

JB Rx.y, jump\_address ; WILL BE INCORRECTLY TAKEN

*Note: The bug is only visible when the bit used for the Jump evaluation is set ('1') for all the bits Rx.15 to Rx.1, or not-set ('0') for Rx.0.*

- Example 5 (BUG NOT VISIBLE)

; Assume Rx.0 is 1 (x any GPR 0..7)

MOVB RxH, any\_value ; Any Byte write instruction on RxH

JB Rx.0, jump\_address ; WILL BE CORRECTLY EXECUTED (TAKEN)

*Note: The bug exist also when the Byte write operation writes into the GPR using indirect addressing mode or memory addressing mode. However, this kind of addressing modes for accessing GPRs are not really expected to be generated by a compiler. This situation includes also the use of PECB/DPECB which destination pointer points into a GPR (that is, data write is performed on a GPR), but this use of PECs is also not expected.*

**Solution**

Workaround:

Include a NOP between any Byte write instruction on a GPR and a Jump-bit instruction on the same GPR but on a bit belonging to a different byte.

*Note: Some more general and easier to implement workarounds can be also defined (however, with unnecessary increase of the code size).*

For example:

- Workaround Alt 1:

Include a NOP between any Byte write instruction and a Jump-bit instruction (on a GPR).

This will cover the case when the Byte write instruction uses indirect addressing mode or memory addressing for accessing GPRs.

- Workaround Alt 2:

Before any jump-bit instruction on a GPR add an "Atomic#1" instruction.

This will cover the case when PECB/DPECB write into the GPR.

Workaround Alt 2 is then the most general, covering all possible cases when the bug can occur, however it is the most expensive in terms of code size and, therefore, is not recommended unless strictly necessary. A possible compromise is to implement "workaround" (or workaround alt 1) in the compiler and in addition to add an assembler checker for workaround alt 2. That is, with each JB/JNB/JBC/JNBS on a GPR warns about possible program flow corruption in case PECB/DPECB are used which destination pointer points into the same GPR used by the Jump-bit instruction (unless they are preceded by an atomic instruction).

## 1.2.4 Cannot Make Write Access to Internal Memory Immediately After Resetting MST\_CLK\_CTRL.CPUH

### ID

WS00008364

### Type

Verification

### Description

When the CPU is at 78 MHz and codes are fetched from Internal RAM, to set clock back to 26 MHz: reset **MST\_CLK\_CTRL.CPUH** and make three write accesses to address 1 BF00<sub>H</sub> (Internal RAM), then verify the written result in this address.

The problem is that the value in address 1 BF00<sub>H</sub> is not the one expected. It seems that we cannot immediately after resetting **MST\_CLK\_CTRL.CPUH**, write accesses cannot be made to the IRAM.

Codes:

```
PUSH  R12
PUSH  R13
MOV   R12,#0FFFEh
EXTP  #03h,#1
AND   02D04h,R12
EXTP  #06h,#1
MOV   R12,03F00h
BFLDL R12,#0Fh,#05h
EXTP  #06h,#1
MOV   03F00h,R12
EXTP  #06h,#1
MOV   R12,03F00h
BFLDL R12,#0F0h,#0A0h
EXTP  #06h,#1
MOV   03F00h,R12
EXTP  #06h,#1
MOV   R12,03F00h
BFLDH R12,#0Fh,#05h
EXTP  #06h,#1
MOV   03F00h,R12
EXTP  #06h,#1
MOV   R12,03F00h
BFLDH R12,#0F0h,#0A0h
EXTP  #06h,#1
MOV   03F00h,R12
POP   R13
POP   R12
```

After these codes the value in address 1 BF00<sub>H</sub> should be A5A5<sub>H</sub>, but it is A505<sub>H</sub>.

### **Solution**

Workaround:

- Add 20 NOPs after switching to 52 MHz
- Add 15 NOPs after switching to 78 MHz.

## **1.2.5 RTC: Power Consumption and Current Peak**

### **ID**

WS00005901

### **Type**

MCU

### **Description**

When the RTC is not in the Isolation mode, the module has a high power consumption and high current peak. Here are the values measured on the Evaluation Board:

Current measurement on E-GOLDRadio:

RTC In Isolation	10.6uA
RTC with Aynchronous Access @ 26 MHz	86.6uA
RTC with Aynchronous Access @ 52 Mhz	241uA
RTC with Aynchronous Access @ 78 Mhz	351uA
RTC with Synchronous Access @ 26 MHz	337uA
RTC with Synchronous Access @ 52 MHz	651uA
RTC with Synchronous Access @ 78 MHz	939uA

Power consumption is too high on VRTC when RTC is in Synchronous mode. Current are above requirements for E-POWERlite.

The impact on overall consumption is limited.

### **Solution**

To reduce power consumption, the RTC register interface has to be:

1. Enabled before the register access.
2. Disabled after the register access.

## **1.2.6 BBRX: The Adjacent Channel Power Estimation Is Biased By Transients**

### **ID**

WS00005865

### **Type**

DSP

Firmware

### **Description**

The bandpass filter (10th order IIR type), adaptively used during phases of high interference by adjacent channels, by its nature has a long settling time, which causes that the transient reaches into the (temporal) estimation window of the adjacent channel power (P\_adj).

This leads to an erroneously too high P\_adj value estimation.

### **Solution**

Workaround:

The wrong estimated adjacent power value P\_adj\_HW is approximately compensated by a firmware patch according to following formula:

$$P\_adj\_FW = P\_adj\_HW + P\_adj\_correction$$

For a delay between BB\_EQON and BB\_RXON of 130us the parameter P\_adj\_correction is different depending on used RF. (e.g.: -200 (SMARTiDC+) and -800 (SMARTiSD)).

Solved in G14\_ER\_STARTUP\_V11.



### **1.2.7 AFE: Noise Signals At Earpiece Buffer Output During Buffer Power Down**

**ID**

WS00005868

**Type**

DSP

**Description**

When earpiece devices in VBRX are powered down ( $VEPPA = 0$ ), a noise signal in band can be observed.

This disappears, if a resistive load of  $< 600$  Ohms is connected between the outputs. This solves the problem with headphones etc., but there may be a problem with the capacitive coupling of the car kit.

This is an inherent phenomenon of the circuit, because in power-down the circuit output is high ohmic and, therefore, on- and off-chip noise signals can couple in path.

**Solution**

Connect resistance (less than 600 Ohms) to earpiece output.

### **1.2.8 I2S: WA0/1 Signal Is Running While TX/RX Is Not Started**

**ID**

WS00005876

**Type**

DSP

**Description**

The problem occurs in Normal/Master mode if both the Tx and Rx paths derive their CLK from the same fractional divider (0/1).

Even if one path (Tx or Rx) is not started (TxStart or RxStart = '0'), the corresponding signals (CLK and WA, Tx) are generated and available on the respective pins.

**Solution**

The FW has to ignore the interrupts from the unused transfer direction.

### **1.2.9 I2S2: RX Path Buggy (polarity bit = 0, Normal Master Mode in 6-Pin Config)**

**ID**

WS00005877

**Type**

DSP

**Description**

The I2S2 RX path does not receive the right values when it's configured as a master in the Normal mode in the 6 pins configuration.

It's due to the polarity bit (POL) in the RXCONF register. If the polarity bit is set to 1, the receive path is OK. If it's set to 0, the first 4 values are wrong in the RX buffer.

**Solution**

Infineon FW does not use the 6 pin configuration.

If the 6 pin configuration is used, the FW must delete first four data words.

### **1.2.10 I2S: RX PCM Does Not Work After Restart**

**ID**

WS00005878

**Type**

DSP

**Description**

The I2Sx does not receive any data in PCM mode after the following sequence:

1. Set all configuration registers and start data transfer
2. Wait for the RX interrupt, data transfer stops because of PCM mode
3. Reset I2SON bit in I2SX.CTRL register
4. Set I2SON bit and start data transfer again.

The first transfer works correctly, but after switching off and on the I2S interface hangs.

**Solution**

SW workaround:

The internal signal can be reset if the I2SControl register is cleared completely when disabling the I2S interface. Use the following sequence:

1. Set all configuration registers and start data transfer
2. Wait for the RX interrupt, data transfer stops because of PCM mode
3. Reset I2SControl register instead of I2SON bit
4. Set I2SControl register and start data transfer again.

### **1.2.11 I2S Doesn't Always Work in Burst Mode**

**ID**

WS00005880

**Type**

DSP

**Description**

The delay (CLK/WA) placed on the Evaluation Board for Normal Mode should also be valid for Burst Mode, but this is not quite correct. Running the Burst Mode correctly implies that the register TxCONF and RxCONF should be set correctly - yet according to the spec these two registers are not relevant for the configuration of the Burst Mode.

**Solution**

-

## **1.2.12 SIM: SIM\_IN IRQ Illegally Occurs After Reset**

### **ID**

WS00005889

### **Type**

MCU

### **Description**

Because of synchronization stages are included for Pad CC\_IN signal the inactive value ('1') produces SIMIN\_INT IRQ two cycles later when reset has already ended. This means that an IRQ is always generated.

This fact results in the following behavior at SIM configuration:

After the first configuration of pad SMC\_IN(=CC\_IN) as input:

- A. A SIM\_IN interrupt is generated if no SIM card is in the simcard holder.
- B. A SIM\_IN interrupt is not generated if there is a SIM card in place.

### **Solution**

After reset, the SIMIN\_INT.IR bit has to be cleared.

No fix planned in hardware.

### **Information**

After reset IRQ-Bit has to be erased.

No fix planned in hardware.

### **1.2.13 SIM: 'Ghost Character' after Retransmission at Parity Error**

**ID**

WS00005893

**Type**

MCU

**Description**

If the SIM block is transmitting and a parity error is detected, after the retransmission of the character a 'ghost character' is observed. This is valid for both the T=0 and Character Modes.

**Solution**

To avoid the observed behavior, the SIM Receive Spacing register SIMTXSPC has to be set to 0003<sub>H</sub>.

### **1.2.14 SSC DSP: Interrupt Bug during Transparent Mode**

**ID**

WS00005897

**Type**

DSP

**Description**

During the Transparent Mode, the SSC does not generate transmit interrupts when writing in the TXB register. Interrupts are generated when a value was transmitted (equal to standard mode).

**Solution**

In the Transparent Mode interrupts are generated as in the standard mode. This means that there are no transmit interrupts generated when writing in the SSCDSP\_TXB register in the Transparent Mode.



### **1.2.15 Corruption of PEC src/dst Segment Address When Overridden by Other PEC/INT**

**ID**

WS00005899

**Type**

MCU

**Description**

The issue occurs on PEC reception transfers from the SIM card. The missed character caused a SIM overrun error that could not be recovered. This issue has also been observed with the camera and display interfaces and is likely to be the source of occasional trace corruption (on ASC1).

When PECs/INTs are requested from the CPU too close to each other, in some special situations the PEC Segment Address of the source/destination pointer can be corrupted. That is, the Segment portion of the address takes the value from the following PEC/INT.

The conditions for this problem are:

1. The CPU is about to process an PEC (first event) but can't do it immediately (due to a data dependency problem). So it delays its processing until the data conflict has been solved (by cancelling the PEC instruction in the Decode stage and re-injected it again later).  
The PEC is however immediately acknowledged by the interrupt controller so that the IC can start arbitrating other events.
2. By the time the first PEC should be re-injected, another PEC or INT is also requested from the CPU (second event). In this case the CPU finishes first the first PEC correctly, however this is not consistently implemented, because the PEC Segment Pointer (PECSNx) that is used is the one corresponding to the second PEC/INT, that is, the address is wrong.

Example:

```
mov 0FCE4h, R9    ; any instruction modifying SRCPx (in this example SRCP1-0FCE4-) with direct or indirect
                  ; addressing mode
mov [R0+], [R1]   ; injected PEC (PEC1 in this example) PEC1 will be canceled and reinjected due to the
                  ; instruction above modifying its SRC pointer. This PEC takes as Segment pointer the one
                  ; corresponding to PEC8 ->WRONG
mov [R0+], [R8]   ; injected PEC (PEC8 in this example)
```

There are 3 situations where a PEC cannot be immediately be processed and is cancelled and reinject again (condition necessary for the bug to happen):

1. The PEC is injected right after an instruction that modifies the PEC Source Pointer (see Example)
2. The PEC Source Pointer points to the PSW register and is injected right after any instruction modifying the PSW flags (almost all instructions modify the PSW flags)
3. The PEC is injected right after an instruction that modifies the CP explicitly (through a SCXT, MOV, etc.).

*Note:*

When PEC Segment Pointers are 0, i.e. only 64 Kbytes regions are addressed by PECs, the problem is not visible. Also, when the PEC Segment Pointers of all the PEC channels (or at least the ones that can happen simultaneously) have the same value, the problem is not visible (in this case, the PECSNx for nodes that trigger an INT instead of a PEC must also have PECSNx programmed to the same value as the one used for the PECs).

*Additional Note:*

Situation 3 was the only one actually seen in the field.

The workaround is implemented on the protocol stack using a Perl script processing the RTOS (OSE) generated assembly code.

It should be implemented in both the RTOS kernel and the C166 C compiler.

## **Solution**

### Workarounds

For situation 1:

Disable all interrupts while modifying src/dst PEC pointers or include these instructions in atomic sequences with an extra instruction after them (a NOP or any instruction not modifying these registers).

Example:

```
atomic #2          ; WORKAROUND
mov 0FCE4h, R9     ; any instruction modifying SRCPx with direct or indirect addressing mode
NOP                ; WORKAROUND (any instruction not modifying SRCPx's)
```

For situation 2:

No real workaround, just do not use PECs with a source pointer to PSW.

For situation 3:

Include any explicit modification of the CP within an atomic-3 sequence (this workaround may be selectively implemented on code sequences where the bug can occur, i.e., at least PECs can occur, different segment pointers are used).

Example:

```
atomic #3          ; WORKAROUND
scxt cp,#new_cp    ; any instruction modifying CP explicitly
...                ; any instr not modifying CP explicitly
...                ; any instr not modifying CP explicitly
```

Notes:

1. A patch (build 234.1.10) is available from Tasking for the A166 assembler and C166 Compiler, version 7.5r5. The new assembler and compiler options make sure the PEC problem does not show up.
2. Compiler Version 8.5 includes the CHECKPECCP switch, which avoids the PEC problem. There is no patch or switch available for Tasking Version 8.0.
3. The effect of the bug is that the SRC and DST pointers can be corrupted. The workaround avoids this situation. It does not need to be applied to modifications of DSTPx pointers (only the SRCPx pointers).

### **1.2.16 Dummy RTC\_INT/RTC\_T14 Interrupts After Reset Phase**

**ID**

WS00005900

**Type**

MCU

**Description**

Due to a incorrect reset implementation some dummy interrupts on RTC\_INT and RTC\_T14 may be generated as soon as the reset is inactive.

Just after the HW reset phase, the RTC block is still in an unstable state because this block is reset afterward by a SW event: so RTC\_INT and RTC\_T14 maybe '0' or '1'.

If RTC\_INT is set to '0', the reset phase is left without detecting any interrupts.

If RTC\_INT is set to '1', the reset phase is left and issues a dummy interrupt to the C166S is issued.

This also applies to RTC\_T14.

**Solution**

At startup (before enabling the interrupts):

Clear interrupt bits T14IR, RTC0IR, RTC1IR, RTC2IR and RTC3IR in RTCISNC register.

Set CLR\_RTCINT bit in RTC\_CTRL register.

### **1.2.17 PM\_INT: the Associated Interrupt Is Missing**

**ID**

WS00005903

**Type**

MCU

**Description**

There is no interrupt vector for the pad PM\_INT.

**Solution**

Workaround: Use S0ASIC as PM\_INT interrupt. The S0ASIC interrupt is generated as soon as wakeup signal is asserted.

If "wakeup on external interrupt" in the SCCU\_HWWAKEUP register is enabled, each time the pad PM\_INT is asserted an interrupt is generated on S0ASIC independent of the frequency of the CPU (32 kHz or 26 MHz).

### **1.2.18 EBU: Not Enough GPIO Can Be Used in VDD\_EBU Domain (Mainly for WP of Flash)**

**ID**

WS00005905

**Type**

MCU

**Description**

In a standard configuration of E-GOLDRadio, only the A23 pin, configured as GPIO, can be used as the write protect pin of a Flash.

If A23 is used as address line, one line from the VDD\_EBU power supply domain is missing, to use the write protect pin of a Flash.

Pads on VDD\_EBU domain are (in a standard configuration) used as their main function and are not available as GPIO.

**Solution**

1. Another GPIOs can be used, when the voltages of VDD\_EBU and VDD\_DIG are almost in the same range. Further investigations have to be done during board design.
2. Use a Level Shifter.

## **1.2.19 I2S: Burst Mode Not Capable Of Quasi-Continuous Transmission**

### **ID**

WS00005909

### **Type**

Top Level

### **Description**

The I2S interface is not capable of quasi-continuous transmission in the burst mode. This is of special importance for the 8 kHz IFX Bluetooth Interface.

The FW needs a certain reaction time ( $>50\mu\text{s}$ ) to reconfigure the RX/TX Start bit. Therefore, a quasi-continuous burst transmission with a 8 kHz (125  $\mu\text{s}$ ) spacing of the WA (sync) signal is not possible.

The problem arises also for higher frequencies.

Clarification of the problem:

In PCM mode (also known as "burst mode"), the start bits are cleared with the TX/RX interrupt; they have to be set again by software to continue the transfer.

The TX interrupt occurs at the rising edge of WA for the last transmitted sample. The RX interrupt occurs when the last bit of the last received sample is latched (in the middle of the last bit period).

In order to achieve a continuous transmission this way, the start bits need to be set again in the time between the TX/RX interrupt and the next WA position. This may be possible for the TX side but is very difficult for the RX side. (Example timing for 8 kHz sample rate: ca. 125  $\mu\text{s}$  for TX, ca 3  $\mu\text{s}$  for RX).

*Note: The "last sample" in this context is the sample in the position before TXINTADDR/RXINTADDR.*

### **Solution**

A FW patch for the slave mode and 8 kSps/s is available.

Do not use master mode or higher frequencies.

HW change not planned.

Workaround:

1. Higher bit clock frequency (slave mode only)

It is possible to gain time for the RX side with a configuration that has some delay between the last bit period and the next WA signal. This is achieved by using a higher bit clock frequency that results in some spare clock periods between the last bit of sample (n) and the beginning of sample (n+1). Such a configuration cannot be set up when the S-GOLDLite is master, but it can be used when the S-GOLDLite is slave.

The BlueMoon Single Cellular PMB8761 as master uses a bit clock of 500 kHz for a sampling rate of 8 kHz, the resulting reaction time after the RX interrupt is about 90  $\mu\text{s}$ .

Generally it is preferable to let S-GOLDLite be the master on the I2S interfaces. (even small clock difference can cause sample under/overflow; slave mode needs more pins on I2S1)

2. Avoid interrupt situation (master or slave mode)

A continuous transfer can also be achieved if the interrupt situation is avoided: while the interface is running, the INTADDR is modified such that the interface logic never reaches the "last sample".

- Use TX interrupt but avoid RX interrupt

The interface could be used with TX interrupts only: at startup and whenever a TX interrupt occurs, the INTADDR registers are set up so that the TX interrupt occurs some samples before the RX interrupt. The sample buffer can be maintained based on RWADDR.

- Avoid all interrupts

With a synchronous DSP loop, e.g. the voiceband scheduler running at 8 kHz, it is also possible to avoid interrupts for both TX and RX by continuously keeping the INTADDR behind the current RWADDR position. Again, the sample buffer can be maintained based on RWADDR.

*Note: The workaround technique "Avoid all interrupts" is used in the current Infineon firmware.*

## **1.2.20 SIM: Auto Powerdown Not SIM Standard Compliant**

### **ID**

WS00005911

### **Type**

MCU

### **Description**

The hardware controlled powerdown function of the SIM interface does not always do a ISO/IEC 7816-3 standard compliant powerdown of the SIM interface.

The hardware controlled powerdown is activated by:

1) Writing a 1 to the SIM\_CTRL.SIMPDWN bit

or

2) Writing a 1 to the SIM\_CTRL.APDWN bit so that any change on input signal SMC\_IN will trigger a hardware controlled powerdown.

The legal powerdown of a SIM has the following sequence:

a) CC\_RST is driven from logic 1 to logic 0

b) CC\_CLK is deactivated

c) CC\_IO is driven to logic 0

d) CC\_VZ deactivates the power to the SIM.

The error is that the CC\_CLK is sometimes:

i) Deactivated before the CC\_RST

or

ii) Deactivated after CC\_CLK, but with a reduced duty cycle (shortened high time).

### **Solution**

Workaround:

Do a software controlled powerdown of the SIM. This means that software deactivates each SIM signal individually. (See SIM specification.)



## **1.2.21     RTC: Alarm Interrupt Generation**

### **ID**

WS00005913

### **Type**

DSP

### **Description**

The spec says:

"One RTC\_CNT cycle after the value in register RTC\_CNT equals the value in RTC\_ALARM, an interrupt request is generated."

But the interrupt is not generated one RTC\_CNT cycle later (i.e., after the value in RTC\_CNT has incremented). It is generated immediately on (or possibly one RTC\_T14CNT cycle after) this condition.

As a consequence, the alarm stays active and cannot be cleared until the next RTC\_CNT cycle, because the alarm condition is still valid.

### **Solution**

Workaround:

Change RTC\_ALARM to another value and clear the Alarm interrupt.

If the old value is to be kept, wait until the next T14 interrupt (RTC\_CNT increased) and then set the RTC\_ALARM register back to the old value.

## **1.2.22 CR110651: Delayed ILLOPA Trap Execution If Caused by Double-Indirect MOV Instruction**

**ID**

WS00005918

**Type**

MCU

**Description**

The illegal word operand access trap (ILLOPA) should be processed whenever a word operand read or write of an odd byte address is started. The class B trap routine expects the stack to contain the instruction following the instruction that caused the trap.

The immediate execution of the ILLOPA trap following the instruction with an illegal odd byte address cannot be guaranteed in case of the double-indirect MOVes (only word MOVes are of interest):

MOV [Rwn+], [Rwm] (opcode D8)

MOV [Rwn], [Rwm+] (opcode E8)

MOV [Rwn], [Rwm] (opcode C8)

Depending on the fill-state of the CPU internal instruction queue, an additional instruction may be executed before the ILLOPA trap is executed. This can be the linear successor of the MOV or any instruction injected at that time, such as an interrupt TRAP.

Depending on the type of instruction that slips in between, the state of the stack may be different from that expected by the class B trap routine.

In any case, the latest stack entry will not contain the IP of the instruction following the instruction that caused the ILLOPA.

**Solution**

-

**1.2.23 CR110652: Delayed ILLBUS Trap Execution When Writing External Bus****ID**

WS00005919

**Type**

MCU

**Description**

The illegal external bus access trap (ILLBUS) should be processed whenever an external instruction fetch, data read, or data write is started and no external bus configuration has been specified. The class B trap routine expects the stack to contain the instruction following the one that caused the trap.

The immediate execution of the ILLBUS trap following the instruction performing the external bus access cannot be guaranteed for instructions that write - but don't read - to the external bus. Instructions that also perform a read access to the external bus (double-indirect MOVes) do not show the problem.

Depending on the fill-state of the CPU-internal instruction queue, an additional instruction may be executed before the ILLBUS trap is executed. This can be the linear successor of the instruction accessing the external bus or any instruction injected at that time, for example, an interrupt TRAP.

Depending on the type of instruction that slips in between, the state of the stack will be different from that expected by the class B trap routine. In any case, the latest stack entry will not contain the IP of the instruction following the one that caused the ILLBUS.

Example:

...

MOV [R12], R1 ; this instruction performs the illegal write access

MOVB RH4, T3 ; this instruction slips unbeaten

TRAP #0Ah ; late TRAP, the stack will not contain the IP value of the MOVB as it should

...

**Solution**

-

## **1.2.24 CR110653: Visible Mode Limitation for Read Accesses on X-Bus**

### **ID**

WS00005920

### **Type**

MCU

### **Description**

Read accesses on the internal X-Bus are not displayed properly on the external bus despite an active visible mode. This happens because the XBC needs an additional cycle to propagate the read data via the write data bus to the external bus.

The data shown in the external bus at the time it would be expected (according to the X-Bus configuration) is an old value and when the right value would be available, it is not really visible on the external pins because the port drivers are already controlled according to the next access.

### **Solution**

Workaround:

Program one or more additional MCTC waitstates than are needed for the data transfer. This gives the XBC the time to propagate the data to the external bus.

At first, the old data is shown on the external bus. Only in the additional cycles are the data correct.

This workaround changes the timing behavior, however this may be acceptable since the Visible Mode is just used for debugging purposes.

### **1.2.25 CR110654: DIP Shows Incorrect Value If Breakpoint Is Inside Atomic/External Sequence**

**ID**

WS00005921

**Type**

MCU

**Description**

The OCDS DIP register should reflect the current program counter value when the CPU goes into the halt mode, for example, due to a hardware breakpoint.

If the hardware breakpoint is inside an atomic sequence, the DIP register will be updated with the IP of the instruction that caused the hardware trigger. However, due to the low priority of the 'CPU halt' respective atomic, all of the protected instructions will be executed before the CPU actually halts.

Example 1:

atomic #4

```
@ mov  R0, #1111h
mov   R0, #2222h
mov   R0, #3333h
mov   R0, #4444h
```

The breakpoint is set on the IP of the first mov. The CPU halts when all four mov instructions have been executed. The DIP register is updated to the IP of the first mov.

Example 2:

atomic #4

```
mov   R0, ONES      ; R_ADR trigger hits -> DIP update
mov   R0, #3333h
mov   R0, #2222h
mov   R0, #1111h
```

The OCDS has been programmed to trigger when address (ONES) is read. Again, the CPU halts when all four mov instructions have been executed. The DIP register is updated to the IP of the first mov.

**Solution**

-> IP breakpoints: Disable placement of IP breakpoints within atomic/extended sequences.

-> Address/data breakpoints: None

## **1.2.26 CR110655: Break-after-Make Does Not Work Correctly in Atomic Scenarios**

**ID**

WS00005922

**Type**

MCU

**Description**

The C166Sv1 OCDS supports break-before-make (BBM) and break-after-make (BAM) for instruction pointer (IP) breakpoints.

For BBM (bit DTREVT.BAM cleared) code execution will stop before the instruction upon which the breakpoint is set is executed.

For BAM (bit DTREVT.BAM set) code execution will stop after the instruction upon which the breakpoint is set is executed.

There is however 1 scenario where the BAM is not performed as it should be: a BAM breakpoint set before an ATOMIC/extended sequence. In this case, the CPU does not go into the Halt mode until the last protected instruction has been executed.

In this scenario, the user also observes the problem according to CPU\_BREAK\_DIP ("DIP shows incorrect value if breakpoint is inside atomic/ext. sequence"), for example, the DIP is not updated to indicate the last instruction that was executed, but it shows the IP of the breakpoint instead.

Example:

A BAM breakpoint is set before an ATOMIC/extended sequence. The CPU does not halt until all the NOP's have been executed. DIP is updated to @

...

@MULR0, R1; hardware breakpoint has been set to this location (BAM)

EXTR#4

NOP

NOP

NOP

NOP

ADDR0, R1

...

**Solution**

Avoid setting BAM breakpoints on instructions followed by atomic/extended sequences.

## **1.2.27 CR110656: Read Address Triggers Cannot Be Generated for All Instruction Reads**

### **ID**

WS00005923

### **Type**

MCU

### **Description**

The C166Sv1 OCDS can be programmed to generate hardware triggers on the address reads bitfield DTREVT.MUX\_R = "11" (R\_ADR)).

Any range of addresses can be programmed using registers DCMPG and DCMPL.

Any C166Sv1 instruction performs up to two operand reads from the address space.

Any implicit GPR (or IDX register for MAC instructions) reads during indirect addressing mode are not taken into account.

Several restrictions apply to the generation of read address triggers in the current implementation:

For each instruction, triggers can be generated for only one operand's read address:

1a. There are instructions performing only one operand read for which read address triggers cannot be generated (see below).

1b. For all instructions performing two operand reads, only one of the two operand's address can be triggered on (see below).

2. (MAC users only): Due to special CoREG addressing scheme, no triggers can be generated on MAC instruction CoSTORE.

\* Details for 1a.

The following instructions perform only one read in the address space (when certain addressing modes are used).

It is not possible to trigger on the related read addresses:

i. ADD(B), ADDC(B), SUB(B), SUBC(B), AND(B), OR(B), XOR(B), CMP(B)

=> applies to addressing modes:

INST Rw/b, #data3

INST reg, #data8/16

ii. CMPD1/2, CMPI1,2

=> applies to addressing modes:

INST Rw, #data4/16

iii. CPL(B), NEG(B)

iv. SCXT reg, #data16

v. MAC instruction CoNOP [IDX\*]

Triggers can be generated for all other instructions performing a single operand read.

*Note: Though the bit-modifying instructions BSET, BCLR, BFLDH, BFLDL perform implicit register reads, the instructions are considered to be write-only.*

Some counter-examples:

- A trigger can be generated on the GPR's address for any "DIVxy Rw" or "EXTxy Rw, #irang"
- MAC: a trigger can be generated for operand address "CoNOP [Rwn\*]" (address of unused read)

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\* Details for 1b.

All instructions performing two operand reads have a syntax corresponding to "INST op1, op2" ;(op1, op2 != immediate values).

Read address triggers are possible only for the address of op2.

Example:

ADD R1, R2

=> It is possible to trigger only on the address of R2.

(MAC users:) This also applies to all MAC instructions "CoXXX op1, op2" ; (op1, op2 != immediate values, CoREG)

### **Solution**

-



## **1.2.28 CR110657: ILLBUS Exception Inside ILLBUS ISR Problem**

**ID**

WS00005924

**Type**

MCU

**Description**

When in the ISR of an ILLBUS trap and another ILLBUS exception is detected due to a Data Read Operation, the core will not resume correctly but waits indefinitely for the second Data Read (unless a Reset is performed).

This situation is very unlikely in a real application. If the External Bus is disabled (the reason for having an ILLBUS), the ILLBUS-ISR will not perform a read operation on the external bus space unless it first enables the bus. Also EPECs/DPECs (the only events that can interrupt a class-B ISR and can lead to the problem) accessing the external bus should not be requested unless the external bus is enabled.

No problem appears if the ISR enables the External Bus before the Data Read Operation that would cause the second ILLBUS trap is executed. Also, no problem appears when the TFR.ILLBUS flag is reset at the beginning of the ISR (and before the Data Read Operation that would cause the second ILLBUS trap is executed).

**Solution**

-

## **1.2.29 CR110764: External Break Events (BREAK Pins) Can Be Lost**

### **ID**

WS00005925

### **Type**

MCU

### **Description**

The C166SV1 is a 2-cycle-machine meaning that a machine cycle has at least two clock cycles.

However, under certain circumstances, for example, insertion of waitstates while fetching instructions or accessing data, a machine cycle takes longer than two clock cycles.

As described in the Integration Manual, the `brk_async_n_i` input must be asserted for at least 2 clock cycles. However, internally this input is sampled/synchronized with machine cycles. So, if there are waitstates, such a short pulse might not be detected.

There is also a similar problem with the `brk_sync_n_i` input. Assuming that it is synchronous to `c166_clk_pos` (as defined in the Integration Manual), it is combinational used in the CPU. However, some of these generated signals are sampled with machine cycle dependant signals and so a short pulse on `brk_sync_n_i` might also be lost.

### **Solution**

Workaround:

Generate pulses (for `brk_async_n_i` and `brk_sync_n_i` inputs) of at least one machine cycle.

To assure a reliable behavior of the Break feature, this workaround can be implemented in the hardware, in the logic responsible for the generation/synchronization of these signals (outside of the C166S core). This logic can make use of the Core output `trc_sel_o` to detect a "machine cycle". One cycle of this signal is identical with one machine cycle. To use the `trc_sel_o` signal the trace bus has to be enabled (`trc_enable_i` '1').

It is suggested to connect `trc_enable_i` to the same signal as `ocds_enable_i`, or if the OCDS can be also enabled by software, to the core output `ocds_enabled_o`.

If the `brk_async_n_i` is used, the `brkout_n_o` output (Activate External Pin action) can be used to deactivate the input signal. The `brk_async_n_i` has already been synchronized when this output is set.

This is not possible for the `brk_sync_n_i` as the `brkout_n_o` is combinational generated in this case.

### **1.2.30 CR111192: Bit-Manipulating Instructions on EXICON/EXISEL Write Wrong Values**

**ID**

WS00005926

**Type**

MCU

**Description**

Bit-manipulating instructions do not work properly on the EXISEL and EXICON registers. Instead of updating just the affected byte, the other bytes are overwritten with wrong data.

Example:

```
MOV R0, #0000h
```

```
EXTR #1
```

```
MOV EXICON, #2222h
```

```
NOP
```

```
EXTR #1
```

BMOV EXICON.1, R0.0 ; expected: EXICON = #2220h, actual (in error): EXICON = #0220h

Affected registers:

EXISEL, EXICON; no other registers are affected by this bug

Affected instructions:

BAND, BCLR, BFLDL, BFLDH, BMOV, BMOVN, BOR, BSET, BXOR, JBC, JNBS

**Solution**

Don't use bit-manipulating instructions that write on these two registers. Instead use word operations (e.g., MOV, AND, OR) and treat them like any other non-bit addressable register.

### 1.2.31 CR112452: Wrong SP Used with SCXT Reg, Mem, and Mem Pointing to SP

**ID**

WS00005927

**Type**

MCU

**Description**

The behavior of the SCXT instruction is as follows:

SCXT reg, mem

(tmp1) = (reg)

(tmp2) = (mem)

(SP) = (SP) - 2

((SP)) = (tmp1)

(reg) = (tmp2)

With this description, in a sequence like

MOV SP, #0FC00h

SCXT R1, 0FC12h ; SP address is 0FE12h

R1 should contain the old SP value 0FC00h. However, the new SP value is written to R1 (0FBFE).

**Solution**

Do not directly access SP in the SCXT instruction, but use an intermediate register.

Example:

MOV R15, SP

SCXT R1, R15

### **1.2.32 RTC: Internal Oscillator Instability**

**ID**

WS00006183

**Type**

Backend

**Description**

The internal oscillator shows a instable jitter in the running CPU:

- D000 is the Jitter measured with the phone on; Sigma = 203.6 Hz peaked noise
- D001 is the Jitter measured with the phone off; Sigma = 78.8 Hz Gaussian noise.

**Solution**

SW workaround:

Perform 32 kHz RTC calibration with the CPU:

- Waiting in a code loop in internal RAM (with no access to external bus)
- Running at 26 MHz pre-scaled by  $2^7$ .

The calibration is run only if more than 16 frames of possible sleep were foreseen and a calibration time out is required. Also, if possible, an equalizer timing offset based calibration algorithm is preferably used when mobile is camped good.

The root cause is due to:

- The very high impedance of 32 kHz oscillator, needed by low power requirement
- Capacitance crosstalk with the address/data/control EBU lines.

### **1.2.33 TEAKlite: Loop (rep) Ends Unexpectedly After Interrupt in PROM Page 1**

**ID**

WS00006189

**Type**

DSP

**Description**

After returning from IR service routine, the first interrupted REP on the extended PROM page (page 1 for E-GOLDRadio) does not continue correctly. The error only occurs one time after reset. All the following REPs cannot be interrupted

**Solution**

During the boot procedure, interrupt the first REP. Because the second REP is not interruptible, the system becomes deterministic.

Solved in G14\_ER\_STARTUP\_V11.

### **1.2.34 DSP: Consecutive Write Operation to Some Registers Fails**

**ID**

WS00006272

**Type**

Top Level

**Description**

When the TEAKlite clock is slower than  $\text{clk\_ms}/2$  (52 MHz) than consecutive write operations to registers clocked by  $\text{clk\_ms}$  (104 MHz) can fail.

**Solution**

Workaround:

Insert NOPs between consecutive writes to registers within a clock domain.

### **1.2.35 Boot Code: Access in External Flash during the BOOTROM**

**ID**

WS00007413

**Type**

MCU

**Description**

The internal boot is proceed from the internal ROM with a 26 MHz frequency, a latched chip select and normal ALE.

This configuration leads to timing violation with some Flashes, when a check is done to know if the external flash contains valid code to be executed or not.

The timing which is violated is coming from the Flash, which needs "ADV/ low when chip select is active (min 10ns)" (according to the Flash specification).

When the boot is directly from the external flash (no bootrom execution), the ALE is automatically configured as extended and there is no timing violation.

**Solution**

No solution.



### **1.2.36 GSM Timer Unit: TFSKIP.SKIPC Bit Has No Effect**

**ID**

WS00007537

**Type**

MCU

**Description**

The TFSKIP.SKIPC bit has no effect of the reset behavior to skip the reset of the current CTDMA counter.

Specified behavior:

If TFSKIP.SKIPC =1 the next time a reset occurs the CTDMA counter value is not set to 0 and continues to be increased over its normal max value.

The TFSKIP.SKIPC is automatically reset internally after skipping the reset of CTDMA counter.

Actual behavior:

The TFSKIP.SKIPC bit has no effect and the reset of the CTDMA counter is processed.

The TFSKIP.SKIPC is automatically reset internally after the reset of CTDMA counter.

**Solution**

Do not use TFSCKIP.SKIPC bit.

Skip functionality have to be handled by SW

### **1.2.37 SCCU X\_BUS Clock at 78 MHz**

**ID**

WS00007621

**Type**

MCU

**Description**

The SCCU must run with 13 MHz to ensure GSM timer synchronization.

At 78 MHz this cannot be achieved due to RST\_CTRL\_STA.SCCUCL bits only allows divisions of the master clock by factor 1; 2; 4, or 8.

The division by 6, which is needed for 78 MHz Mode to setup 13 MHz, is not supported.

This means, the Stand-By Mode cannot be entered in the 78 MHz mode.

**Solution**

Use 26 MHz or 5 MHz before entering the Stand-By Mode.

### **1.2.38     RTC: Internal Oscillator Instability**

**ID**

WS00006183

**Type**

Backend

**Description**

The internal Oscillator shows a instable Jitter behavior, by running CPU.

**Solution**

-

### **1.2.39 GSM Timer Unit: TFSKIP.SKIPC Bit Has No Effect**

**ID**

WS00007537

**Type**

DSP

**Description**

The TFSKIP.SKIPC bit has no effect of the reset behavior to skip the reset of the current CTDMA counter.

Specified behavior:

If TFSKIP.SKIPC = 1, the next time a reset occurs the CTDMA counter value is not set to 0 and continues to be increased over its normal max value.

The TFSKIP.SKIPC is automatically reset internally after skipping the reset of CTDMA counter.

Actual behavior:

The TFSKIP.SKIPC bit has no effect and the reset of the CTDMA counter is processed.

The TFSKIP.SKIPC is automatically reset internally after the reset of CTDMA counter.

**Solution**

Do not use TFSKIP.SKIPC bit, the Skip functionality has to be handled by SW.

## **1.2.40 SSC: Phase Error When High Baudrate Is Set**

### **ID**

WS00008268

### **Type**

Evaluation Board

### **Description**

The SSCs (PD Bus SSC and DSP SSC) support a phase error detection, as defined in the spec:

A phase error is detected if the phase error detection is enabled and the receive data change is within 1 module clock cycle before and 2 module clock cycles after the receive edge of the shift clock SSCx\_SCLK.

If the shift clock SSCx\_SCLK is generated by a fractional divider, a phase error may occur due to the fractional divider jitter.

The fractional divider generates clock low/high phases of different lengths. In case of a short phase, the difference between the receive edge and the transmit edge of the shift clock may be too small as compared to the module clock period and thus violates the condition given above.

This may happen, for example, if an SSC is master and receives data. It always happens in this case if the shift clock frequency is between 1/4 and 1/2 of the module clock frequency.

This may also happen if the shift clock frequency is between 1/8 and 1/4 of the module clock frequency.

In the latter case, the critical clock and data edges sometimes occur logically at the same time and it depends on pad and board timings whether a phase error is detected or not.

No problem occurs if the shift clock frequency is exactly 1/2, 1/4 or 1/8 or below 1/8 of the module clock frequency.

Also no problem occurs if the master SSC does not generate a fractional divider jitter or if the receiving SSC does not support/not enable phase error detection.

If a receiving SSC outside of the E-GOLDradio supports phase error detection and it is enabled, the detection may follow other rules as compared to the E-GOLDradio SSCs. In this case, other frequencies may be critical as in the test case given above.

### **Solution**

-

## **1.2.41 EBU: Improve Timings for EBU Interface**

### **ID**

WS00007486

### **Type**

Physical Implementation

### **Description**

EBU Timings need to be improved to fulfil specified target values.

### **Solution**

To be done in a future version of E-GOLDRadio.



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